WJOULWATT

JW5712/JW5712A

Ultra-Low Quiescent Current Synchronous Step-Down Converter

Preliminary Specifications Subject to Change without Notice

DESCRIPTION

The JW®5712/A provides an ultra-low power solution for products powered by either one rechargeable Li-lon batteries, Li-primary battery chemistries such as Li-MnO2 or two to three cell alkaline batteries. The output voltage of the buck converter is set with three VSEL pins between 0.7V and 3.3V. JW5712/A features low output ripple voltage and low noise with a small output capacitor. Once the input voltage comes close to the output voltage the device enters the 100% duty cycle operation mode to prevent an increase of output ripple voltage. In this operation mode the device stops switching and turns the high-side MOSFET switch on.

JW5712/A is available in 8 ball WLCSP 1.6mmx0.9mm package, which provide a compact solution with minimal external components.

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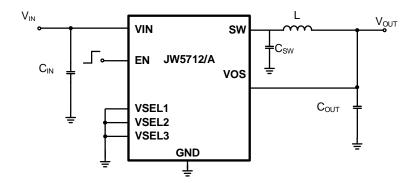
FEATURES

- 2.3V to 5.5V operating input range
- Output current
 JW5712/A: 600mA, peak to 800mA
- Typical 360nA quiescent current
- Up to 90% efficiency with load current>15uA
- Low output ripple voltage
- 8 selectable output voltages JW5712: 1.2V to 3.3V JW5712A: 0.7V to 3.1V
- Output short protection
- Output voltage discharge
- Automatic transition to 100% duty cycle operation
- Thermal protection
- Available in WLCSP8 package

APPLICATIONS

- IOT
- Wearable and Personal Electronics
- Health Monitoring and Medical Accessories
- Industrial Metering
- Energy Harvesting

TYPICAL APPLICATION



ORDER INFORMATION

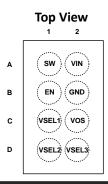
DEVICE ¹⁾	PACKAGE	TOP MARKING ²⁾	ENVIRONMENTAL ³⁾
JW5712WLCSPC#TR	WLCSP1.575x0.875-8	JWWI □	Green
JW5712AWLCSPC#TR	WLCSP1.575x0.875-8	JWNH□ YW□□□	Green

Note:



3) All Joulwatt products are packaged with Pb-free and Halogen-free materials and compliant to RoHS standards.

PIN CONFIGURATION



ABSOLUTE MAXIMUM RATING1)

0.3V to 6V
0.3V to 6V
150°C
260°C
65°C to +150°C
±2kV
±500V

RECOMMENDED OPERATING CONDITIONS³⁾

Input Voltage VIN	2.3V to 5.5V
JW5712/A Output Current lout	600mA
Junction Temperature Range	40°C to 125°C

THERMAL PERFORMANCE⁴⁾ θ_{JA} θ_{Jctop}

WLCSP8 103 1°C/W

Note:

1) Exceeding these ratings may damage the device. These stress ratings do not imply function operation of the device at any other conditions beyond those indicated under RECOMMENDED OPERATING CONDITIONS.

- 2) The JW5712/JW5712A includes thermal protection that is intended to protect the device in overload conditions. Continuous operation over the specified absolute maximum operating junction temperature may damage the device.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB

ELECTRICAL CHARACTERISTICS

VIN=3.6V, T_A = -40 $^{\circ}$ C to 85 $^{\circ}$ C, ty	pical values	are at T_A = 25 C , Unles	ss otherw	ise state	d.	
ltem	Symbol	Conditions	Min.	Тур.	Max.	Unit
$V_{\rm IN}$ Under-voltage Lockout Threshold	V _{IN_MIN}	V _{IN} rising	2.09	2.16	2.25	V
V _{IN} Under-voltage Lockout Hysteresis	V _{IN_MIN_HYST}			100		mV
Shutdown Supply Current	I _{SD}	V _{EN} =0V		40	300	nA
Supply Current	IQ	device not switching		360	600	nA
Supply Current		device switching ⁵⁾		460	800	nA
High Side MOSFET On-resistance	R _{DS(ON)T}	V _{IN} =3.6V, I _{OUT} =50mA		380	470	mΩ
Low Side MOSFET On-resistance	$R_{DS(ON)B}$	V _{IN} =3.6V, I _{OUT} =50mA		250	450	mΩ
High Side MOSFET Switch Current Limit	I _{LIM_Peak}	3V≤V _{IN} ≤5.5V	1.3	1.5	1.7	А
Low Side MOSFET Switch Current Limit	I _{LIM_Valley}	3V≤V _{IN} ≤5.5V	1	1.2	1.4	Α
Output Discharge Resistance	R _{DISCH}	EN=GND, Ivos=-10mA into VOS pin		30	65	Ω
Bias Current Into VOS Pin	I _{IN_VOS}	V _{OUT} =2V, EN=V _{IN}		100	200	nA
Output Voltage Accuracy	V_{OUT}	V _{OUT} =1.8V	-2%		+2	%
5)		V _{OUT} =1.8V,CCM operation, I _{OUT} =0~600mA		0.01		%/mA
utput Voltage Load Regulation ⁵⁾		V _{OUT} =1.8V, I _{OUT} =100mA, 2.3V≤V _{IN} ≤5.5V		0.002		%/mA
Output Voltage Line Regulation ⁵⁾		V _{OUT} =1.8V, I _{OUT} =100mA, 2.3V≤V _{IN} ≤5.5V		0.1		%/V
Minimum On Time	T _{ON_MIN}	I _{OUT} =0mA, V _{OUT} =2V		500		ns
Minimum Off Time ⁵⁾	T _{OFF_MIN}			80		ns
Soft-start Delay Time ⁵⁾	T _{SS_DLY}	From EN=low to high until device starts switching		8		ms
Soft-start Time ⁵⁾	T _{ss}	2.3V≤V _{IN} ≤5.5V, EN=V _{IN}		830		us
High Side MOSFET Switch Current Limit During Soft-start ⁵⁾	I LIM_Peak_SS	EN=low to high		150		mA

VIN=3.6V, T_A = -40 $\mathcal C$ to 85 $\mathcal C$, typical values are at T_A = 25 $\mathcal C$, Unless otherwise stated. Unit Item **Symbol Conditions** Min. Тур. Max. Low Side MOSFET Switch Current 128 mΑ EN=low to high LIM_Valley_SS Limit During Soft-start⁵⁾ 2.3V≤V_{IN}≤5.5V, ٧ Input High Threshold V_{IH} 1.2 EN, VSELs pins 2.3V ≤V_{IN}≤5.5V, Input Low Threshold V V_{IL} 0.4 EN, VSELs pins Input Pin Bias Current I_{IN} 10 nΑ Rising V_{IN}, 100% Auto 100% Mode Leave Detection mode is left with 350 V_{TH_100+} 150 250 mV Threshold $V_{IN}=V_{OUT}+V_{TH_100+}$ Falling V_{IN}, 100% Auto 100% Mode Enter Detection mode is entered with 85 200 290 mV $V_{TH\ 100-}$ Threshold $V_{IN}=V_{OUT}+V_{TH_100}$ ° C Thermal Shutdown⁵⁾ $\mathsf{T}_{\mathsf{TSD}}$ 150 Thermal Shutdown Hysteresis⁵⁾ $\mathsf{T}_{\mathsf{TSD_HYST}}$ 20 °C

Note:

5) Guaranteed by design.

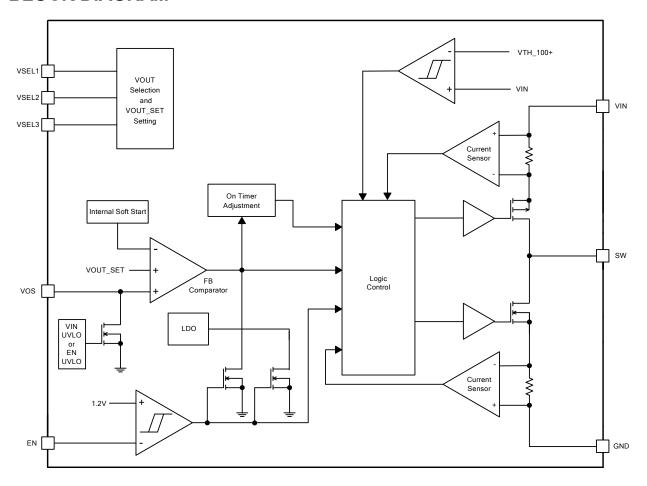
Table 1. Output Voltage Setting

JW5712 VOUT/V	JW5712A VOUT/V	VSEL3	VSEL2	VSEL1
1.2	0.7	0	0	0
1.5	1	0	0	1
1.8	1.3	0	1	0
2.1	0.75	0	1	1
2.5	1.9	1	0	0
2.8	1.05	1	0	1
3.0	2.9	1	1	0
3.3	3.1	1	1	1

PIN DESCRIPTION

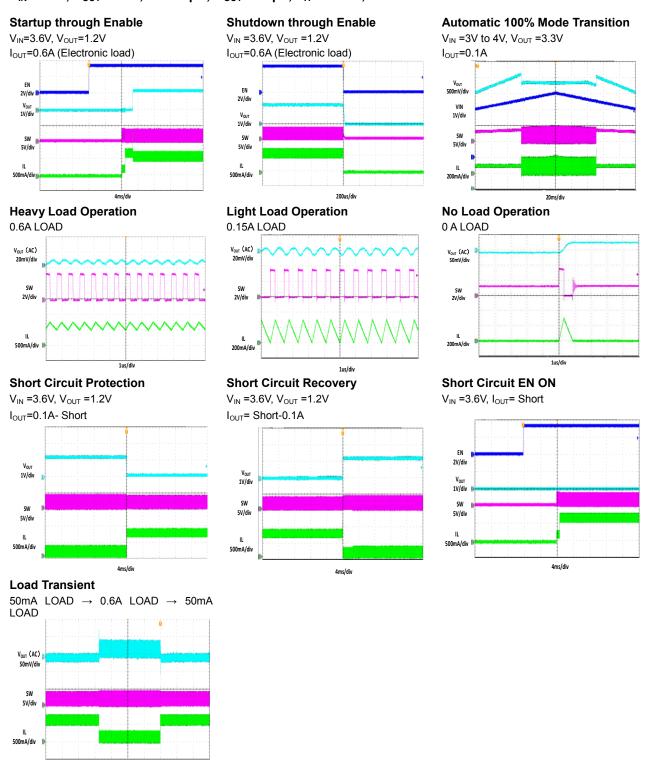
Pin	Name	Description	
A1	SW	Switch pin which is connected to the internal MOSFET switches. Connect inductor to this	
AI		terminal.	
A2	VIN	VIN power supply pin. Connect this pin close to the VIN terminal of the input capacitor. A	
AZ		minimum of 4.7μF ceramic capacitor from this pin to GND is required.	
B1	EN	Enable pin. Drive EN pin high to turn on the regulator and low to turn off the regulator.	
B2	GND	Ground pin.	
C1	VSEL1		
D1	VSEL2	Output voltage selection pins.	
D2	VSEL3		
C2	VOS	Feedback pin for the internal feedback divider network and regulation loop. Connect this	
62		pin directly to the output capacitor with a short trace.	

BLOCK DIAGRAM



TYPICAL PERFORMANCE CHARACTERISTICS

 V_{IN} =3.6V, V_{OUT} = 1.2V, L = 2.2 μ H, C_{OUT} = 10 μ F, T_A = +25°C, unless otherwise noted



TYPICAL PERFORMANCE CHARACTERISTICS

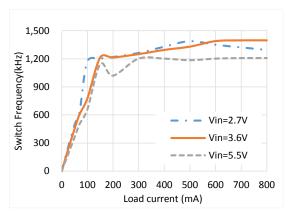


Figure1: Frequency VS Load Current V_{OUT} =1.2V

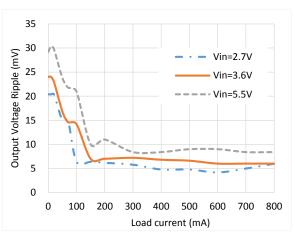


Figure3: Vout Ripple voltage VS Load Current

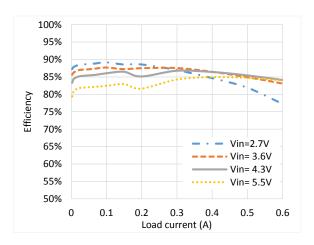


Figure5: Efficiency vs Load Current (V_{OUT} =1.2V, L=2.2µH)

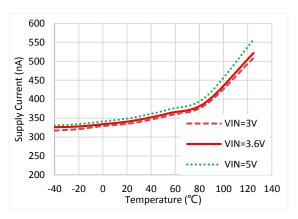


Figure2: IQ VS Temperature Device not switching

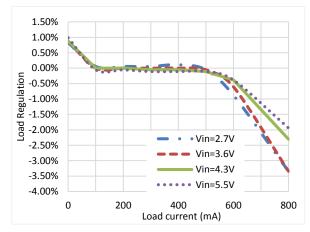


Figure4: Load Regulation vs Load Current

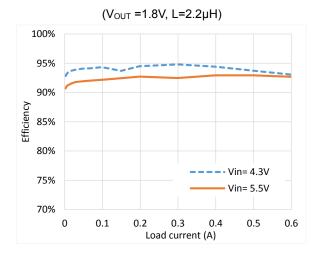


Figure6: Efficiency vs Load Current (V_{OUT} =3.3V, L=2.2µH)

FUNCTIONAL DESCRIPTION

JW5712/A is an ultra-low power synchronous step-down regulator. It regulates input voltages from 2.3V to 5.5V down to an output voltage range from 0.7V to 3.3V with ultra low quiescent current consumption (360nA typical) and is capable of supplying to 600mAof continue load current and 800mA of peak load current.

Shut-Down Mode

JW5712/A shuts down when voltage at EN pin is driven below 0.3V. The entire regulator is off and the supply current consumed by JW5712/A drops below 40nA.

Output Voltage Selection

JW5712/A does not require an external resistor divider network to program the output voltage. The device integrates a high impedance feedback resistor divider network which is programmed by VSEL1-3 pins. JW5712/A supports an output voltage ranges of 0.7V to 3.3V. The output voltage can be changed during operation and supports simple dynamic output voltage scaling. The output voltage is programmed according to **Table 1**.

Soft Start

The JW5712/A has an internal soft-start function to minimize input voltage drop during start-up. This allows the operation from high impedance battery cells. Once the device is enabled the device starts switching after a typical delay time of 8ms, then the soft-start time of typical 830µs begins with a reduced current limit of typical 128 mA. When this time passed by the device enters full current limit operation. This allows a smooth start-up and the device can start into full load current. Furthermore, larger output capacitors impact the start-up behavior of the DC/DC converter especially when the output voltage

does not reach its nominal value after the typical soft-start time of 830µs has passed.

Output Current Run-Away Protection

JW5712/JW5712A ingrates a current limit on the high side as well on the low side MOSFETs to protect the device against overload or short circuit conditions. The peak current in the switches is monitored cycle by cycle. If the high side MOSFET current limit is reached, the high side MOSFET is turned off and the low side MOSFET is turned on until the current decreases below the low side MOSFET current limit. When the voltage of FB decrease below to 0.4V, the current limit will decrease to 1/2 current limit.

Output Discharge

The device provides automatic output voltage discharge once it is disabled. This feature prevents residual charge voltage on the output capacitor, which may impact proper power up of the system connected to the converter. The discharge circuit at VOS pin becomes active once the EN pin is pulled to low or the input voltage drops below UVLO comparator threshold.

100% Duty Cycle Operation Mode

Once the input voltage comes close to the output voltage, the DC/DC converter stops switching and enters 100% duty cycle operation mode and the output voltage V_{OUT} tracks the input voltage V_{IN} . Once the input voltage VIN falls below the 100% mode enter threshold, V_{TH_100-} , the DC/DC regulator is turned off, switching stops and therefore no output voltage ripple is generated. The voltage difference between V_{IN} and V_{OUT} is the voltage drop across the power inductor and the internal high side

MOSFET switch. Once the input voltage increases and trips the 100% mode exit threshold, V_{TH_100+} , the DC/DC regulator turns on and starts switching again.

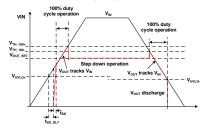


Figure 7. Automatic 100% Mode Transition

Thermal Protection

When the temperature of the JW5712/A rises above 150°C, it is forced into thermal shut-down. Only when core temperature drops below 130°C can the regulator becomes active again.

APPLICATION INFORMATION

Input Capacitor

The input capacitor is used to supply the AC input current to the step-down converter and maintaining the DC input voltage. The ripple current through the input capacitor can be calculated by:

$$I_{CIN} = I_{OUT} * \sqrt{\frac{V_{OUT}}{V_{IN}} * \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}$$

where I_{OUT} is the load current, V_{OUT} is the output voltage, V_{IN} is the input voltage.

Thus the input capacitor can be calculated by the following equation when the input ripple voltage is determined.

$$C_{IN} = \frac{I_{OUT}}{f_{SW}*\Delta V_{IN}}*\frac{V_{OUT}}{V_{IN}}*\left(1 - \frac{V_{OUT}}{V_{IN}}\right)$$

where C_{IN} is the input capacitance value, f_{sw} is the switching frequency, ΔV_{IN} is the input ripple voltage.

The input capacitor can be electrolytic, tantalum or ceramic. To minimizing the potential noise, a small X5R or X7R ceramic capacitor, i.e. 0.1µF, should be placed as close to the IC as possible when using electrolytic capacitors.

A $4.7\mu F$ ceramic capacitor is recommended in typical application.

Output Capacitor

The output capacitor is required to maintain the DC output voltage, and the capacitance value determines the output ripple voltage. The output voltage ripple can be calculated by:

$$\Delta V_{OUT} = \frac{V_{OUT}}{f_{SW}*L}*\left(1 - \frac{V_{OUT}}{V_{IN}}\right)*\left(R_{ESR} + \frac{1}{8*f_{SW}*C_{OUT}}\right)$$

where C_{OUT} is the output capacitance value and R_{ESR} is the equivalent series resistance value of the output capacitor.

The output capacitor can be low ESR electrolytic, tantalum or ceramic, which lower ESR capacitors get lower output ripple voltage. The output capacitors also affect the system stability and transient response, and a 10µF ceramic capacitor is recommended in typical application.

Inductor

The recommended power inductor is 2.2uH and inductor saturation current rating choose follow over current protection design consideration. In applications, it needs to select an inductor with the low DCR to provide good performance and efficiency.

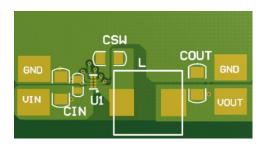
External Csw Capacitor

The C_{SW} capacitor must be added between SW pin and GND pin. A 1nF low ESR ceramic capacitor is recommended to connected between the SW pin and GND pin. Keep the C_{SW} as close as close to SW pin and GND pin.

PCB Layout Note

For minimum noise problem and best operating performance, the PCB is preferred to following the guidelines as reference.

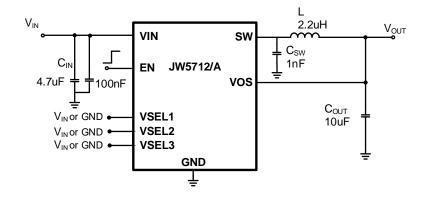
- Place the input decoupling capacitor as close to JW5712 and JW5712A (VIN pin and GND pin) as possible to eliminate noise at the input pin. The loop area formed by input capacitor and GND must be minimized.
- 2. Keep the C_{SW} as close as close to SW pin and GND pin.
- 3. Keep the switching node SW short to prevent excessive capacitive coupling
- Make V_{IN}, V_{OUT} and ground bus connections as wide as possible. This reduces any voltage drops on the input or output paths of the converter and maximizes efficiency.



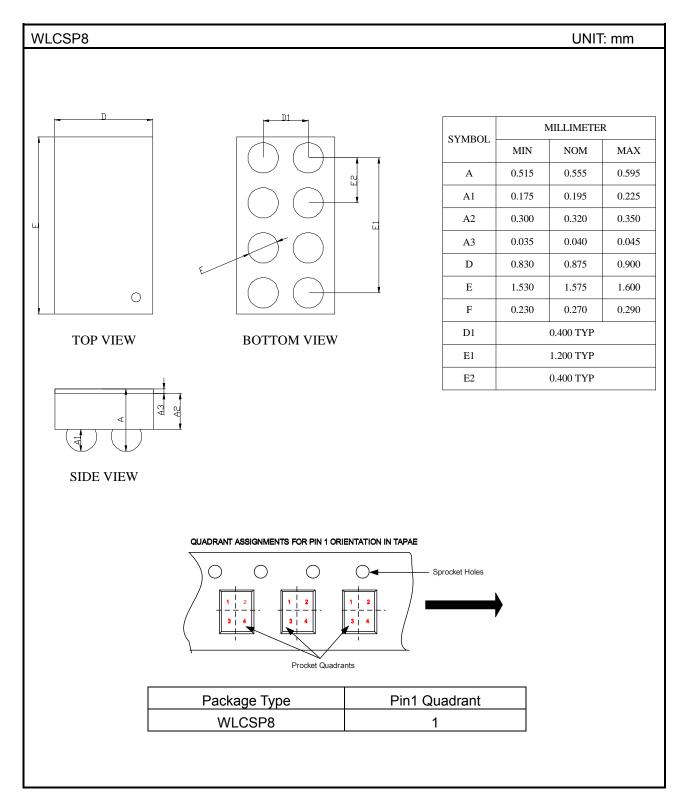
REFERENCE DESIGN

V_{IN}: 2.7V~5.5V

I_{OUT}: 0~0.6A



PACKAGE OUTLINE



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